

Amirkabir Journal of Mechanical Engineering

Amirkabir J. Mech Eng., 53(special issue 2) (2021) 255-258 DOI: 10.22060/mej.2020.16893.6468

Attitude Stabilization of Quadrotor Using Adaptive Fuzzy Proportional Integral Derivative Controller

H. Chehardoli*1, A. Ghasemi2, S. Fardrahnama2

¹Department of Mechanical Engineering, Ayatollah Boroujerdi University, Boroujerd, Iran

ABSTRACT: Quadrotor is an unmanned aerial robot from multi-rotor drones group that has high maneuverability, vertical take-off, and landing and stationary flight capabilities. In the most practical applications, the quadrotor system is subjected to external disturbance forces due to wind and unbalanced weight or inertia of the payload. To maintain balance and hold the position, attitude stabilization of the quadrotor is necessary for the presence of disturbances and unbalanced forces. Using conventional controllers with constant gains is not very efficient to eliminate variable disturbances that affect quadrotor motion in different conditions. In this paper, an adaptive fuzzy proportional integral derivative controller is designed for quadrotor attitude stabilization in which controller gains are regulated continuously based on the adaptive laws and the fuzzy inference system. The performance of the proposed controller is examined in the disturbance rejection test and is compared to the conventional proportional integral derivative controller. Also, the performance of the proposed controller is approved by hardware in the loop experimental tests using a 3 degree of freedom pilot platform. The experimental results will show the effectiveness of the adaptive fuzzy proportional integral derivative controller compared with the conventional proportional integral derivative controller.

Review History:

Received: Aug. 13, 2019 Revised: Dec. 27, 2019 Accepted: Jan. 26. 2019 Available Online: Feb. 02, 2020

Keywords:

Quadrotor

Attitude control

Adaptive fuzzy

Hardware in loop

External disturbance

1. INTRODUCTION

In recent decades, the stability analysis and control design of various kinds of Unmanned Aerial Vehicles (UAVs) have received much attention. The UAVs are widely used in several tasks such as visual acquisition, surveillance, exploration, and disaster assistance in urban circumstances [1]. Consequently, it is studied for different kinds of UAVs such as fixed-wing airplanes, helicopters, and also quadrotors. The quadrotor is the most popular kind of multi-rotor UAVs due to its simple mechanics, high maneuverability, and performing stable stationary flights.

The motion control of a quadrotor is a challenging problem due to its nonlinear under-actuated dynamics. Especially, controlling the lateral motion of a quadrotor is the main problem since it is associated with an under-actuated subsystem of the quadrotor dynamics. Several approaches are presented to control design and stability analysis of quadrotors and some strategies have been developed to solve the path following problems [2, 3].

The model-based methods presented to control the design of quadrotors involve two major drawbacks. They need fast and heavy computation units due to their complicated and time-consuming control laws. On the other hand, they are restricted for autonomous long-range applications where the imposed communication delay with stationary control unit disrupts real-time operation of the quadrotor [4]. Motivated by previous studies, in this paper, fuzzy logic is employed to design an adaptive fuzzy-PID controller for nonlinear under-

actuated dynamics of the quadrotor. To achieve a robust performance against external disturbances, all PID gains are updated individually. Moreover, a compensator is added to control structure to weaken the estimation errors of the adaptive fuzzy-PID controller. Performance of the proposed adaptive fuzzy-PID controller is compared with conventional PID controller which shows significant improvement in tracking accuracy.

2. DYNAMIC MODELING AND CONTROL DESIGN

The Quadrotor has a simple mechanical structure consisting of a symmetric cross-shaped rigid body and four electrical rotors which are located at the end of cross arms. The schematic model of a quadrotor is depicted in Fig. 1. The trust force (T) and the aerodynamic moment (Q) are calculated as follows [5]

$$T = b\omega^2, \qquad Q = d\omega^2 \tag{1}$$

where b and d are constant positive values and ω is the angular velocity. The translational dynamics of a quadrotor is given from the Newton approach [5]

$$\begin{pmatrix} \ddot{X} \\ \ddot{Y} \\ \ddot{Z} \end{pmatrix} = \begin{pmatrix} U_1 \left(c_{\phi} s_{\theta} c_{\psi} + s_{\phi} s_{\psi} \right) / m \\ U_1 \left(c_{\phi} s_{\theta} s_{\psi} - s_{\phi} c_{\psi} \right) / m \\ U_1 c_{\phi} c_{\theta} / m - g \end{pmatrix}$$
 (2)

The rotational dynamics is presented as follows [

*Corresponding author's email: h.chehardoli@abru.ac.ir

Copyrights for this article are retained by the author(s) with publishing rights granted to Amirkabir University Press. The content of this article Copyrights for this article are retained by the author(s) with publishing rights granted to the terms and conditions of the Creative Commons Attribution 4.0 International (CC-BY-NC 4.0) License. For more information,

² Department of Mechanical Engineering, Faculty of Engineering, North Tehran Branch, Islamic Azad University, Tehran, Iran

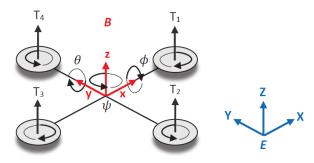


Fig. 1. Schematic model of a quadrotor





Fig. 2. Experimental setup

$$\begin{pmatrix} \ddot{\phi} \\ \ddot{\theta} \\ \ddot{\psi} \end{pmatrix} = \begin{pmatrix} \frac{U_2 - J_r \dot{\theta} \omega_r + (I_{yy} - I_{zz}) \dot{\psi} \dot{\theta}}{I_{xx}} \\ \frac{U_3 + J_r \dot{\phi} \omega_r + (I_{zz} - I_{xx}) \dot{\psi} \dot{\phi}}{I_{xx}} \\ \frac{U_4 + (I_{xx} - I_{yy}) \dot{\theta} \dot{\phi}}{I_{zz}} \end{pmatrix}$$
 (3)

The quadrotor is an under-actuated system because it has six degrees of freedom but only four actual inputs. As a result, only four DOFs can be controlled independently. Eq. can be represented as follows

$$\begin{pmatrix} \ddot{\phi} \\ \ddot{\theta} \\ \ddot{\psi} \end{pmatrix} = \begin{pmatrix} \frac{-J_r \theta \omega_r + (I_{yy} - I_{zz})\dot{\psi}\theta}{I_{xx}} \\ \frac{J_r \dot{\phi}\omega_r + (I_{zz} - I_{xx})\dot{\psi}\dot{\phi}}{I_{xx}} \\ \frac{I_{xx}}{I_{zz}} \end{pmatrix} + \begin{pmatrix} \frac{1}{I_{xx}} & 0 & 0 \\ 0 & \frac{1}{I_{yy}} & 0 \\ 0 & 0 & \frac{1}{I_{yy}} \end{pmatrix} \begin{pmatrix} U_1 \\ U_2 \\ U_3 \end{pmatrix} + \begin{pmatrix} d_{\phi} \\ d_{\theta} \\ d_{\psi} \end{pmatrix} = \mathbf{F} + \mathbf{G}\mathbf{U} + \mathbf{D}$$

$$(4)$$

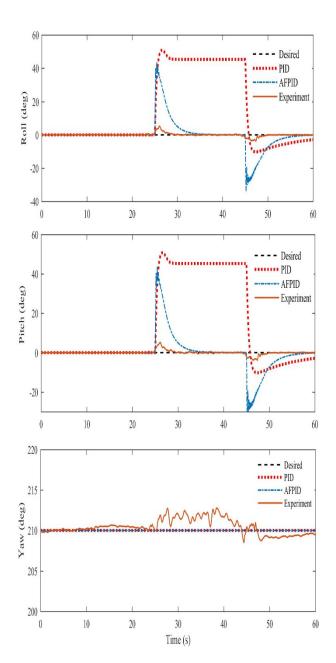


Fig. 3. Stabilization performance in presence of external disturbance

The following feedback linearization controller is introduced to yield three independent dynamical equations

$$U = -F/G + \tau \tag{5}$$

For each of the resultant second-order equations, the same control procedure is designed. These equations are in the following form

$$\ddot{x} = f(x, \dot{x}) + g(x, \dot{x})u + d, \qquad y = x \tag{6}$$

where f(.) and g(.) are nonlinear bounded functions. By defining the tracking error as $\mathbf{e} = y - y_r$, we will have

$$\ddot{\mathbf{e}} + \mathbf{k}^T \mathbf{e} = g(\mathbf{x})(u - u^*) + \mathbf{d},$$

$$u^* = \frac{-1}{g(\mathbf{x})} \left(f(\mathbf{x}) - \ddot{y}_r + \mathbf{k}^T \mathbf{e} \right)$$
(7)

For the above system, an adaptive fuzzy-PID controller is designed as $\tau = \tau_{PID} + v$. The term v is a H_{∞} compensator defined by $v = -\mathbf{B}^T \mathbf{P} \mathbf{e} / \lambda$ where λ is a positive value and \mathbf{P} is a positive definite matrix calculated by the following equation

$$\mathbf{A}T\mathbf{P} + \mathbf{P}\mathbf{A} + \mathbf{Q} - \frac{2}{\lambda}\mathbf{P}\mathbf{B}\mathbf{B}^{T}\mathbf{P} + \frac{1}{\rho^{2}}\mathbf{P}\mathbf{B}\mathbf{B}^{T}\mathbf{P} = \mathbf{0}, \quad 2\rho^{2} \ge \lambda$$
 (8)

The control input $\mathbf{\tau}_{PID}(\xi \mid \theta) = \mathbf{\theta}^T \xi(\mathbf{e})$ is defined where $\dot{\mathbf{e}} = [K_P, K_I, K_D], \mathbf{\hat{1}}(\mathbf{e}) = [\mathbf{e}, \mathbf{e}_I, \mathbf{e}_D]^T, \mathbf{e}_I = \int \mathbf{e}(t) dt$ and $\mathbf{e}_D = \dot{\mathbf{e}}(t)$. All gains K_P, K_I and K_D are estimated by a Zero-order Takagi-Sugeno fuzzy system.

3. EXPERIMENTAL STUDY

To validate the proposed approach, a 3DOF experimental setup is investigated as shown in Fig. 2.

Fig. 3 depicts the stabilization performance of roll, pitch and yaw angles in presence of external disturbance. As this Figure illustrates, the proposed adaptive fuzzy-PID controller has a significantly better performance compared with the conventional PID controller.

4. CONCLUSIONS

In this paper, the angular stabilization problem of quadrotors was studied. An adaptive fuzzy-PID controller was presented to assure the asymptotic tracking of yaw, roll, and pitch angles in presence of nonlinear uncertain dynamics and external disturbance. Compared with the previous studies, in this paper, all PID control gains were updated individually which provided better performance in tracking the desired signals. Experimental results were provided to show the effectiveness of the proposed approach.

5. REFERENCES

- [1] [1] K.P. Valavanis, G.J. Vachtsevanos, Handbook of unmanned aerial vehicles, Springer, 2015.
- [2] [2] R. Mahony, V. Kumar, P. Corke, Multirotor aerial vehicles: Modeling, estimation, and control of quadrotor, IEEE robotics & automation magazine, 19(3) (2012) 20-32.
- [3] [3] A. Tayebi, S. McGilvray, Attitude stabilization of a VTOL quadrotor aircraft, IEEE Transactions on control systems technology, 14(3) (2006) 562-571.
- [4] [4] H. Bolandi, M. Rezaei, R. Mohsenipour, H. Nemati, S.M. Smailzadeh, Attitude control of a quadrotor with optimized PID controller, (2013).
- [5] [5] E. Kayacan, R. Maslim, Type-2 fuzzy logic trajectory tracking control of quadrotor VTOL aircraft with elliptic membership functions, IEEE/ASME Transactions on Mechatronics, 22(1) (2016) 339-348.

HOW TO CITE THIS ARTICLE

H. Chehardoli, A. Ghasemi, S. Fardrahnama, Attitude Stabilization of Quadrotor Using Adaptive Fuzzy Proportional Integral Derivative Controller. Amirkabir J. Mech Eng., 53(special issue 2) (2021) 255-258.

DOI: 10.22060/mej.2020.16893.6468



This Page intentionally left blank